

Identification for Control: A ν -gap metric Approach

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Introduction

Questions Addressed

Given Two feedback loops : model-controller ('designed' loop), unknown true plant-controller ('achieved' loop)

1. Given frequency response samples of the true plant, **find** a model that approximates the plant in a suitable metric such that the designed loop \approx the achieved loop for 'any' controller.
2. Given a bounded power excitation r , **characterise** the difference in the response of the designed and the achieved closed-loop to excitation r for 'any' controller.

Notation

G_i normalised right graph symbol for
 $P_i = N_i M_i^{-1}$ ($G_i = \begin{bmatrix} N_i \\ M_i \end{bmatrix}$ is inner)

\tilde{G}_i normalised left graph symbol for
 $P_i = \tilde{M}_i^{-1} \tilde{N}_i$ ($\tilde{G}_i = [-\tilde{M}_i \ \tilde{N}_i]$
 is co-inner)

$\eta(P)$ number of unstable poles of P

wno winding number; for $X, X^{-1} \in \mathcal{RL}_\infty$,
 $\text{wno det}(X) = \eta(X^{-1}) - \eta(X)$

$I(P_0, P_1)$ wno det $(G_1^* G_0)$

$\mathcal{R}^{m \times n}$ Finite dimensional LSI systems
 (m o/p, n i/p)

The ν -gap metric

The ν -gap between two plants P_0 and P_1 may be defined by

$$\begin{aligned} \delta_\nu(P_0, P_1) &:= \inf_{\substack{Q, Q^{-1} \in \mathcal{L}_\infty \\ \text{wno } \det(Q)=0}} \|G_0 - G_1 Q\|_\infty \\ &:= \sup_\omega \kappa(P_0, P_1)(e^{j\omega}) \text{ when} \\ &\quad I(P_0, P_1) = 0 \end{aligned}$$

The *pointwise chordal distance* $\kappa(P_0, P_1)$ defined by

$$\begin{aligned} \kappa(P_0, P_1)(e^{j\omega}) &:= \\ &\bar{\sigma} \left((I + P_0 P_0^*)^{-\frac{1}{2}} (P_0 - P_1) (I + P_1^* P_1)^{-\frac{1}{2}} \right) (e^{j\omega}) \\ &= \bar{\sigma} (\tilde{G}_2 G_1) (e^{j\omega}) \end{aligned}$$

The ν -gap metric (cont'd)

Define closed loop transfer function for a plant P_i and controller C by

$$H(P_i, C) = \begin{bmatrix} P_i \\ I \end{bmatrix} (I - CP_i)^{-1} \begin{bmatrix} -C & I \end{bmatrix}$$

Fact: For a plant P_0 , model P_1 and a controller C , at any point $e^{j\omega}$,

$$\begin{aligned} \kappa(P_0, P_1) &\leq \bar{\sigma}(H(P_0, C) - H(P_1, C)) \text{ and} \\ \bar{\sigma}(H(P_0, C) - H(P_1, C)) &\leq \alpha \kappa(P_0, P_1) \end{aligned} \quad (1)$$

where $\alpha = \bar{\sigma}(H(P_1, C))\bar{\sigma}(H(P_0, C))$.

Hence, $\delta_\nu(\text{plant, model})$ is small

\Leftrightarrow difference in closed loop behaviour is small for any 'reasonable' controller

Identification Problem

Given: Frequency response samples of plant P_0 ,

$$P_0(e^{j\omega_i}), i = 1, 2, \dots, m$$

A ‘sensible’ identification problem:

Find P_1 such that

$$\begin{aligned} \inf_{Q, Q^{-1} \in \mathcal{L}_\infty} \max_i \bar{\sigma}(G_0 - G_1 Q)(e^{j\omega_i}) \\ = \max_i \kappa(P_0, P_1)(e^{j\omega_i}) \end{aligned}$$

is minimised; subject to a smoothness constraint on P_1 and $I(P_0, P_1) = 0$.

Strategy for an Approximate Solution

- Formulate an equivalent problem, in terms of $P_0(e^{j\omega_i})$ instead of $G_0(e^{j\omega_i})$;
- Choose a parameterised model set \mathcal{S} , solve

$$\min_{P \in \mathcal{S}} \max_i \kappa(P_0, P)(e^{j\omega_i})$$

subject to some smoothness constraint on P . Let P_1 be the solution.

- Given a model P_1 and a controller C s.t. $\eta(H(P_0, C)) = 0$, find another model P_2 which satisfies $I(P_0, P_2) = 0$ and minimises $\max_{\omega} \kappa(P_1, P_2)$. Then

$$\max_i \kappa(P_0, P_2)(e^{j\omega_i}) \leq \max_i \kappa(P_0, P_1) + \max_{\omega} \kappa(P_1, P_2)$$

An Equivalent Problem Formulation

Lemma 1 Given $P_0, P_1 \in \mathcal{RL}_\infty$, $Q, Q^{-1} \in \mathcal{L}_\infty$,
 $\exists \hat{Q}, F$ such that $\hat{Q}, \hat{Q}^{-1}, F \in \mathcal{L}_\infty$,

$$\bar{\sigma}(G_0 - G_1 Q)(e^{j\omega}) = \bar{\sigma}(F - G_1 \hat{Q})(e^{j\omega}) \forall \omega$$

and at any ω , complex matrix $F(e^{j\omega})$ can be written as a function of point frequency response matrix $P_0(e^{j\omega})$.

For a SISO system, we can choose F as

$$F(e^{j\omega_i}) = \left[\frac{P_0(e^{j\omega_i})}{\sqrt{1 + |P_0(e^{j\omega_i})|^2}} \quad \frac{1}{\sqrt{1 + |P_0(e^{j\omega_i})|^2}} \right]^T$$

Choice of Model Set (SISO case)

Let $S_n = \text{span} \{1, z, \dots, z^{n-1}\}$ and

$$S_{1,2} = \{f : f = [f_1 \ f_2]^T, f_1 \in S_{n_1}, f_2 \in S_{n_2}\}$$

For $f \in S_{1,2}$, the constraints

$$\max_i \bar{\sigma}(G_0 - fQ)(e^{j\omega_i}) < \lambda \text{ and}$$

$\|f'\|_\infty < \lambda$ are affine in the parameters of f .

Chordal Distance Minimisation

Initialisation: Set $k = 1$, $\hat{Q}_0 = 1$.

Step A: Solve LMI optimisation

$$\min_{f_k \in \mathcal{S}_{1,2}} \max_{i \in [1,m]} \left\{ \bar{\sigma}(F(e^{j\omega_i}) - f_k(e^{j\omega_i})\hat{Q}_{k-1}(e^{j\omega_i})), \right. \\ \left. \alpha \|f'_k\|_\infty \right\}$$

where α is a user specified weight; $F(e^{j\omega_i})$ derived from $P_0(e^{j\omega_i})$. Let \hat{f}_k be the solution.

Step B: Solve

$$\min_{Q_k(e^{j\omega_i})} \max_{i \in [1,m]} \bar{\sigma}(F(e^{j\omega_i}) - \hat{f}_k(e^{j\omega_i})Q_k(e^{j\omega_i}))$$

Let $\hat{Q}_k(e^{j\omega_i})$ be the solution. If $\max_i \bar{\sigma}(F(e^{j\omega_i}) - \hat{f}_k(e^{j\omega_i})\hat{Q}_k(e^{j\omega_i}))$ is less than a specified tolerance, stop; otherwise set $k := k + 1$ and go back to step A.

For $\hat{f}_k = [\hat{f}_1 \hat{f}_2]^T$, the model obtained is $P_1 = \hat{f}_1 \hat{f}_2^{-1}$.

Winding Number Adjustment - 1

Idea (*Vinnicombe, 1993*):

Given P_1, P_2, C satisfying

$$\kappa(P_1, P_2)(e^{j\omega}) \bar{\sigma}(H(P_1, C))(e^{j\omega}) < 1 \forall \omega,$$

$$\eta(H(P_2, C)) = \eta(H(P_1, C)) + I(P_2, P_1)$$

\Rightarrow Given a controller C which stabilises plant P_0 and a model P_1 such that

$$\eta(H(P_1, C)) = k$$

find another model P_2 such that

$$I(P_2, P_1) \leq -k$$

$$\Downarrow \text{ if } \kappa(P_1, P_2)(e^{j\omega}) \bar{\sigma}(H(P_1, C))(e^{j\omega}) < 1 \forall \omega$$

$$\eta(H(P_2, C)) = 0$$

$$\Downarrow \text{ if } \kappa(P_2, P_0)(e^{j\omega}) \bar{\sigma}(H(P_2, C))(e^{j\omega}) < 1 \forall \omega$$

$$I(P_2, P_0) = 0 \quad (\text{since } \eta(H(P_0, C)) = 0)$$

Winding Number Adjustment - 2

Solution: Given P_1 , solution of the chordal distance approximation problem and a controller C that stabilises P_0 , Solve

$$\inf_{\substack{P \in \mathcal{R} \\ I(P, P_1) \leq -k}} \sup_{\omega} \kappa(P_1, P)(e^{j\omega})$$

where $k = \eta(H(P_1, C))$.

The solution to above problem may be characterised by solution of a Hankel norm approximation problem. If P_2 is a solution,

$$\max_i \kappa(P_0, P_2)(e^{j\omega_i}) \leq \max_i \kappa(P_0, P_1) + \max_{\omega} \kappa(P_1, P_2)$$

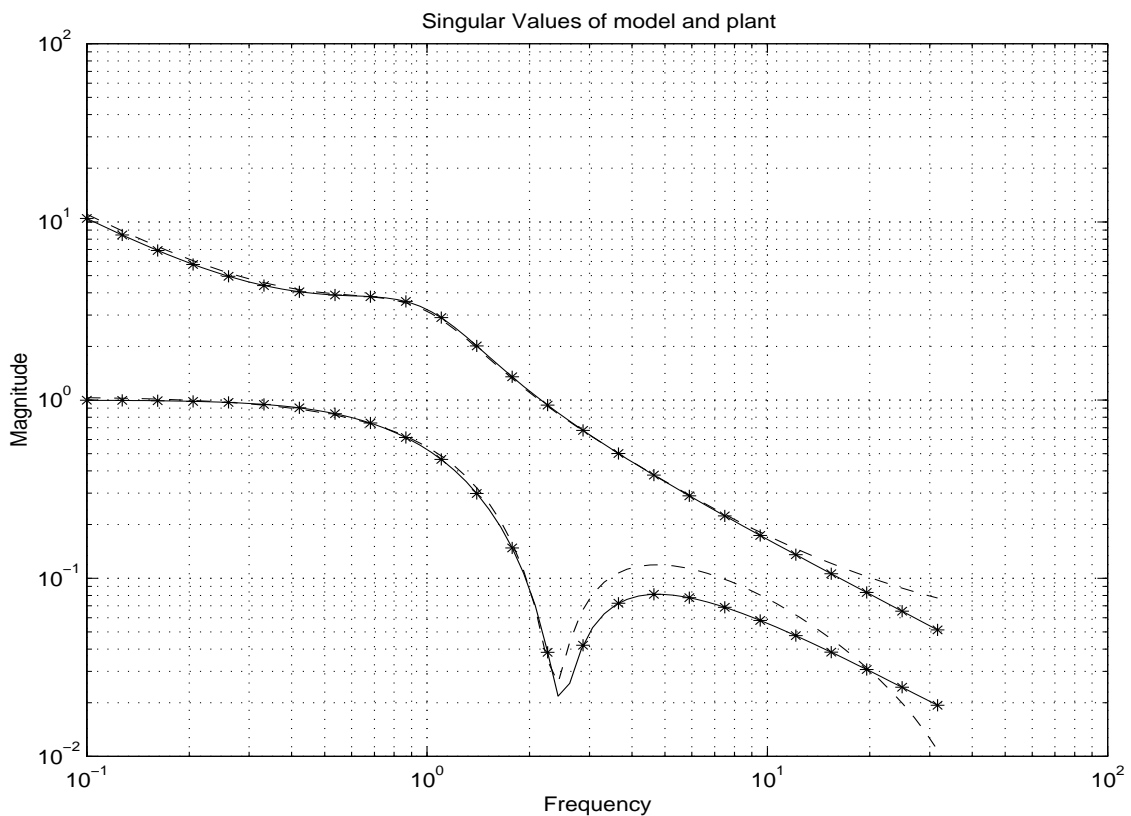
If $\eta(H(P_0, 0)) = \eta(H(P_1, 0)) = 0$, and $\max_i \kappa(P_0, P_1)(e^{j\omega_i})$ is ‘small’, winding number adjustment is not required (i.e. P_1 is our final model.).

Simulation Example

$$P_0(s) = \begin{bmatrix} \frac{1}{s+2} & \frac{1}{s} \\ \frac{1}{s-1} & \frac{3}{s^2+s+1} \end{bmatrix}$$

25 non-uniformly spaced freq. response samples. $\alpha = .0010$, order = 3 (each term).

Singular Value Plots



$$\max_i \kappa(\text{plant}, \text{model})(e^{j\omega_i}) = 0.06 \approx \delta_\nu(\text{plant}, \text{model})$$

The Second Question

- Given a bounded power excitation r , **characterise** the difference in the response of the designed and the achieved closed-loop to excitation r for ‘any’ controller.

$$\kappa(P_0, P_1) \leq \bar{\sigma}(H(P_0, C) - H(P_1, C)) \text{ and}$$
$$\bar{\sigma}(H(P_0, C) - H(P_1, C)) \leq \alpha \kappa(P_0, P_1)$$

where $\alpha = \bar{\sigma}(H(P_1, C))\bar{\sigma}(H(P_0, C))$.

Does a similar result hold pointwise in the set of persistent signals?

Bounded Power Signals

- Let $R_u(\tau) = \lim_{N \rightarrow \infty} \frac{1}{N} \sum_{t=0}^{N-1} u(t - \tau) u^T(t)$.

- $\mathcal{S}^n = \{u \mid u \in l_{\infty}^n(\mathbb{Z}), u(t) = 0 \forall t < 0,$

$$R_u(\tau) \text{ exists } \forall \tau, \phi_u(\omega) := \sum_{\tau=-\infty}^{\infty} R_u(\tau) e^{-j\tau\omega}$$

exists for all ω

- *Power spectrum* $\phi_u(\omega)$ need not be bounded and may contain impulses.

- $\|f\|_{\mathcal{S}} := \sqrt{\text{trace } R_f(0)} = \sqrt{\frac{1}{2\pi} \int_{-\pi}^{\pi} \text{trace } \phi_f(\omega) d\omega}$.

- Subset of interest:

$$\mathcal{S}^{n'} := \{v \mid v \in \mathcal{S}^n, \phi_v = \phi_{\tilde{v}} I_{n \times n} \text{ with } \tilde{v} \in \mathcal{S}^1\}$$

A New Measure of Distance over $\mathcal{R}^{m \times n}$

- Let $\mathcal{R}_2^{m \times n} := \{(P_1, P_2) : P_1, P_2 \in \mathcal{R}^{m \times n}\}$

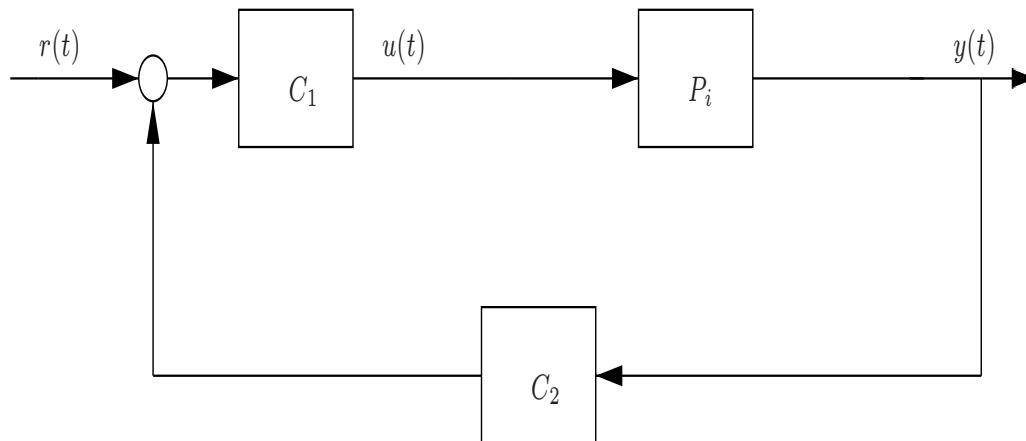
Define $\delta_r : \mathcal{R}_2^{m \times n} \times \mathcal{S}^{n'} \rightarrow \mathbb{R}$

$$\begin{aligned} \delta_r(P_1, P_2, r) &:= \frac{\left\{ \frac{1}{2\pi} \int_{-\pi}^{\pi} \text{trace} (\tilde{G}_2 G_1) \phi_r (\tilde{G}_2 G_1)^* d\omega \right\}^{\frac{1}{2}}}{\|r\|_S} \\ &= \frac{\|\tilde{G}_2 G_1 r\|_S}{\|r\|_S} \end{aligned}$$

- $0 \leq \delta_r(P_1, P_2, r) \leq 1$.
- For any $r_0 \in \mathcal{S}^{n'}$ such that $\phi_{r_0} \neq 0 \forall \omega \in [-\pi, \pi]$, $\delta_r(P_1, P_2, r_0)$ is a metric.
- For $n = 1$,

$$\sup_{r \in \mathcal{S}^{1'}} \delta_r(P_1, P_2, r) = \sup_{\omega} \kappa(P_1, P_2) =: \delta_{\mathcal{L}_2}(P_1, P_2)$$

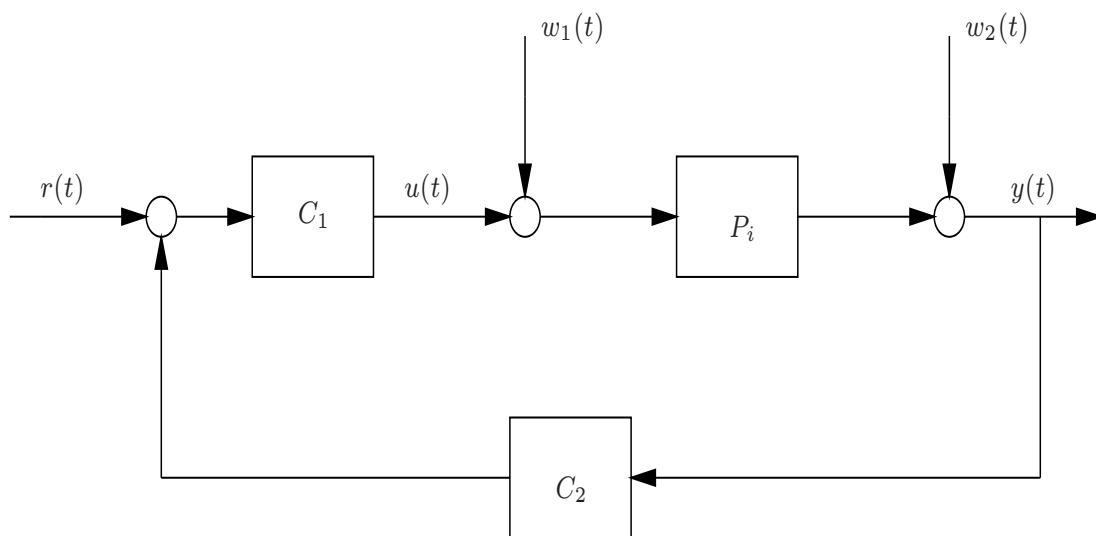
Closed Loop Set-up



- C_1 is such that $\underline{\sigma}(C_1)(e^{j\omega}) > 0$.
- $T(P_i, C_1, C_2) =$ transfer function from r to $\begin{bmatrix} y \\ u \end{bmatrix}$.
- $b(P_i, C) = \|H(P_i, C)\|_{\infty}^{-1}$ ('generalised robust stability margin'; larger the better).

Bounds on δ_r

Data Generating System



- **Given:** Measured data $z := \begin{bmatrix} y \\ u \end{bmatrix}$,
power $\| r \|_s$.

- r, w_i such that

$$\lim_{N \rightarrow \infty} \frac{1}{N} \sum_{t=0}^{N-1} (r(t) w_i^T(t - \tau)) = 0 \quad \forall \tau, i = 1, 2$$

Example 1

- $P_1(z) = \frac{2(z+1)}{z-0.6}$, $P_2(z) = \frac{(z+1)^2}{(z^2-0.6z+1.2)}$.

- $\delta_\nu(P_1, P_2) = 0.64$.

- If $\phi_{r_0} = f(e^{j\omega})f(e^{-j\omega})$, $f(z) = \frac{0.01z}{z-0.99}$,

$$\delta_r(P_1, P_2, r_0) = 0.0417.$$

- $C_1 = \frac{0.025(z+1)}{(z-1)}$, $C_2 = -1$

↓

$$b(P_1, C) = 0.408, b(P_2, C) = 0.401.$$

Example 2

- $P_3(z) = \frac{z-1}{z-0.99}$, $P_4(z) = \frac{z-1}{z-1.01}$.
- $\delta_\nu(P_1, P_2) = 1$.
- If $\phi_{r_0} = f(e^{j\omega})f(e^{-j\omega})$, $f(z) = 1$,
 $\delta_r(P_1, P_2, r_0) = 0.055$.
- $C_1 = 1$, $C_2 = -1$
 \Downarrow
 $b(P_1, C) = b(P_2, C) = 0.7071$.

Summary

- A new algorithm for frequency domain identification; identification strategy to minimise the difference in the closed loop behaviour between the model and the plant for *any* controller.
- A new measure of distance for systems under bounded power excitation; time domain characterisation of difference in closed loop behaviour under persistent reference/disturbances.

Finally,

Related Work (*by other people*)

- Bombois, Gevers, Scorletti, ECC'99: Given $\epsilon, X, \hat{\theta}$,

$$\Phi = \{\theta : (\hat{\theta} - \theta)^T X (\hat{\theta} - \theta) < \epsilon\}$$

Solve $\min_{\theta \in \Phi} \kappa(P_{\hat{\theta}}, P_{\theta})$.

- Steele, Vinnicombe, submitted to CDC'01: Given $\hat{P}, \{y_N, u_N\}, \epsilon, \delta$,

$$\mathcal{P} = \{P : \|y_N - Pu_N\|_q \leq \delta, \delta_\nu(P, \hat{P}) \leq \epsilon\}$$

Is \mathcal{P} non-empty?